

# Solutions

EECS 20, Section 2, Fall 2015

Midterm #2, November 5, 2015

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## Problem 1

Select one or more correct choices for each question. NO PARTIAL CREDITS.

1.1 (5 Points) A system represented by a set of LCCDEs is causal and LTI if and only if

- a. All auxiliary conditions are set to 0.
- b. The set of LCCDEs have the same number of terms corresponding to the inputs and the outputs.
- c. The condition of the initial rest is satisfied.
- d. The LCCDEs have removable poles.

1.2 (5 Points) Suppose a set of LCCDEs represent an LTI system, and the Frequency

Response is given by  $H(\omega) = \frac{1+2e^{-i2\omega}}{4+e^{-i\omega}-3e^{-i3\omega}}$ . Then the Difference Equation satisfied by the system is

- a.  $y(n) + 2y(n-2) = 4x(n) + x(n-1) - 3x(n-3)$ .
- b.  $4y(n) + y(n-1) - 3y(n-3) = x(n) + 2x(n-2)$ .
- c.  $y(n) + 4y(n-1) + 2y(n-3) = x(n) + 2x(n-2)$ .
- d. None of the above.

1.3 (5 Points) For a given discrete-time system we know that input of  $\delta(n)$  produces the output of  $\delta(n+1) + \delta(n)$ . Then,

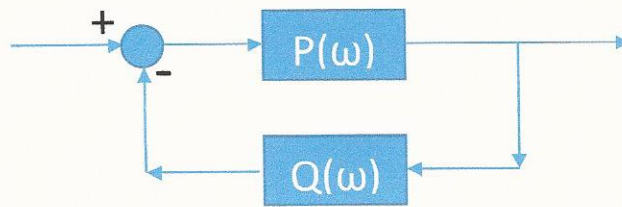
- a. The system cannot be causal.
- b. The system cannot be memoryless.
- c. The system can be causal, but not necessarily.
- d. The system can be memoryless, but not necessarily.

1.4 (5 Points) A system is defined to be BIBO stable if and only if

- a. The system output is always bounded.
- b. The system only accepts bounded input.
- c. The system has unbounded output only when the input is unbounded.
- d. None of the above.

## Problem 2

Consider the discrete-time feedback system shown below:



2.1 (5 Points) Corresponding to the Frequency Response  $P(\omega)$ , the Impulse Response is given by  $p(n) = (0.2)^n u(n)$ ,  $n \in \mathbf{Z}$ . Find  $P(\omega)$ .

$$\begin{aligned} P(\omega) &= \sum_{n \in \mathbf{Z}} p(n) e^{-in\omega} = \sum_{n \in \mathbf{Z}} (0.2)^n u(n) e^{-in\omega} \\ &= \sum_{n=0}^{\infty} (0.2)^n e^{-in\omega} = \sum_{n=0}^{\infty} (0.2 e^{-i\omega})^n \\ &= \frac{1}{1 - 0.2 e^{-i\omega}} \end{aligned}$$

2.2 (5 Points) The subsystem with the Frequency Response  $Q(\omega)$  satisfies the input-output relationship  $y(n) = 0.5 * (x(n) + x(n-1))$ ,  $n \in \mathbf{Z}$ . Find  $Q(\omega)$ .

Let  $x(n) = e^{in\omega}$ . Then,

$$\begin{aligned} y(n) &= 0.5 (e^{in\omega} + e^{i(n-1)\omega}) \\ &= 0.5 e^{in\omega} (1 + e^{-i\omega}) \end{aligned}$$

Since we expect  $y(n) = e^{in\omega} Q(\omega)$ ,

$$Q(\omega) = 0.5(1 + e^{-i\omega})$$

2.3 (10 Points) Let  $h(n), n \in \mathbb{Z}$  be the Impulse Response of the overall feedback system. Find  $h(n)$ .

By Black's Equation,  $H(\omega) = \frac{1}{1 - 0.2e^{-i\omega}} \cdot \frac{1}{1 + \frac{0.5(1 + e^{-i\omega})}{1 - 0.2e^{-i\omega}}}$

Simplifying,  $H(\omega) = \frac{1}{1 - 0.2e^{-i\omega} + 0.5 + 0.5e^{-i\omega}} = \frac{1}{1.5 + 0.3e^{-i\omega}} = \frac{2}{3} \frac{1}{1 + 0.2e^{-i\omega}}$

Since  $\sum_{n=0}^{\infty} \alpha^n = \frac{1}{1-\alpha}$ ,  $h(n) = \frac{2}{3} (-0.2)^n u(n)$ .

2.4 Fill in each blank with LPF, HPF, Comb Filter, or Notch Filter.

a. (5 Points) The subsystem  $P(\omega)$  is a LPF.

b. (5 Points) The subsystem  $Q(\omega)$  is a LPF. (since  $Q$  corresponds to two-point moving average)

c. (5 Points) The overall feedback system is a HPF.

For a. & c., recall that impulse response  $\alpha^n u(n)$  corresponds to a LPF for  $\alpha \in (0, 1)$ , and a HPF for  $\alpha \in (-1, 0)$ .

2.5 (5 Points) Is the overall feedback system BIBO stable? Justify your answer.

Since the system is LTI, BIBO stable

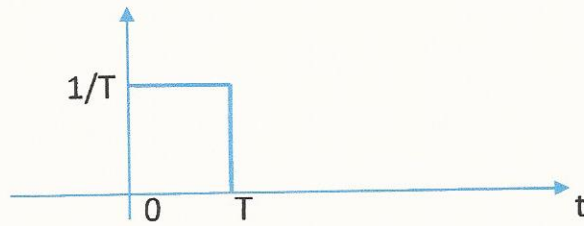
iff  $\sum_{n \in \mathbb{Z}} |h(n)| < \infty$

Here, we have  $\sum_{n \in \mathbb{Z}} |h(n)| = \frac{2}{3} \sum_{n \in \mathbb{Z}} |(-0.2)^n| = \frac{2}{3} \sum_{n \in \mathbb{Z}} (0.2)^n = \frac{2}{3} \cdot \frac{1}{1-0.2} = \frac{5}{6} < \infty$

Hence, BIBO stable.

### Problem 3

Suppose the Impulse Response  $h(t)$  of a continuous-time LTI system is as shown below:



3.1 (10 Points) Find the Frequency Response  $H(\omega)$ .

$$\begin{aligned} H(\omega) &= \int_{-\infty}^{\infty} h(t) e^{-i\omega t} dt = \frac{1}{T} \int_0^T e^{-i\omega t} dt \\ &= \frac{1}{T} \frac{e^{-i\omega t}}{-i\omega} \Big|_0^T = \frac{1}{T} \cdot \frac{1}{i\omega} (1 - e^{-i\omega T}) \\ &= e^{-i\omega T/2} \left( \frac{e^{i\omega T/2} - e^{-i\omega T/2}}{2i} \right) \cdot \frac{1}{\omega T/2} \\ &= e^{-i\omega T/2} \frac{\sin \omega T/2}{\omega T/2} = e^{-i\omega T/2} \cdot \text{sinc}(\omega T/2) \end{aligned}$$

3.2 (10 Points) For a general input signal  $x(t)$ , how is the output  $y(t)$  is related to the input  $x(t)$ ?

$$\begin{aligned} y(t) &= h * x(t) \\ &= \int_{-\infty}^{\infty} h(\tau) x(t-\tau) d\tau \\ &= \frac{1}{T} \int_0^T x(t-\tau) d\tau \end{aligned}$$

Let  $t-\tau = u$ . Then,  $d\tau = -du$ ,  $\tau=0 \Rightarrow u=t$  &  $\tau=T \Rightarrow u=t-T$

$$y(t) = -\frac{1}{T} \int_t^{t-T} x(u) du = \frac{1}{T} \int_{t-T}^t x(u) du$$

i.e.  $y(t)$  is average of  $x(t)$  over the previous  $T$  units of time (over  $t-T$  to  $t$ ).

3.3 (5 Points) Is the system causal? Justify your answer.

Causal. Since  $y(t) = \frac{1}{T} \int_{t-T}^t x(u) du$ ,  $y(t)$

only depends on the present & past values of  $x(t)$ .

3.4 (5 Points) Is the system memoryless? Justify your answer.

Not memoryless. Since  $y(t) = \frac{1}{T} \int_{t-T}^t x(u) du$ ,

$y(t)$  clearly depends on the past values of  $x(t)$ .

3.5 (10 Points) For the input  $x(t) = e^{i\omega t}$ ,  $t \in \mathbb{R}$ , we find that the output is always 0 at the frequencies  $\omega = \pm k\pi$ ,  $k = 1, 2, 3, \dots$ . Find the value of  $T$ .

For input  $e^{j\omega t}$ , output is  $e^{j\omega t} H(\omega)$ .

So, output will be 0 at zero's of  $H(\omega)$ .

Now,  $H(\omega) = e^{-i\omega T/2} \text{sinc}(\omega T/2)$

Hence,  $|H(\omega)| = 0$  iff  $\omega T/2 = \pm m\pi$  for some  $m \in \{1, 2, 3, \dots\}$ .

Since we want  $|H(\omega)| = 0 \forall \pm k\pi$ ,  $k \in \{1, 2, 3, \dots\}$ ,

the possible values of  $T$  are  $\{2, 4, 6, \dots\}$ .

(Smallest possible value of  $T$  is 2).