

EE40 Midterm 1 Solutions  
Fall 2013  
Mitchell Kline

$$\text{par}(R_1, R_2) := \frac{R_1 \cdot R_2}{R_1 + R_2}$$

Problem 1 a

$$V_{oc} := \alpha \cdot V_x \quad I_{sc} := \beta \cdot V_x \quad R_{th} := \frac{V_{oc}}{I_{sc}} \rightarrow \frac{\alpha}{\beta}$$

The above parameters specify a Thevenin source with voltage source  $V_{oc}$  and resistance  $R_{th}$ . We need to find an equivalent circuit that has voltage  $V_x$  rather than  $V_{oc}$  on the left hand side. There are many possible solutions, but the simplest one is a voltage divider.

For a voltage divider with series resistance  $R_1$  and load  $R_2$ , the Thevenin parameters are:

$$V_{oc} := \frac{R_2}{R_1 + R_2} \cdot V_x \quad I_{sc} := \frac{V_x}{R_1}$$

Equating these with the above, we can find  $\alpha$  and  $\beta$  by inspection.

$$\text{Given} \quad \alpha = \frac{R_2}{R_1 + R_2} \quad \beta = \frac{1}{R_1}$$

$$\text{Find}(R_1, R_2) \text{ simplify} \rightarrow \begin{bmatrix} \frac{1}{\beta} \\ \alpha \\ -\frac{\alpha}{\beta \cdot (\alpha - 1)} \end{bmatrix}$$

With three resistors in a T,  $R_1$ ,  $R_2$ ,  $R_3$

$$V_{oc} := \frac{R_2}{R_1 + R_2} \cdot V_x \quad I_{sc} := \frac{\frac{\text{par}(R_2, R_3)}{R_1 + \text{par}(R_2, R_3)} \cdot V_x}{R_3} \text{ simplify} \rightarrow \frac{R_2 \cdot V_x}{R_1 \cdot R_2 + R_1 \cdot R_3 + R_2 \cdot R_3}$$

$$\text{Given} \quad \alpha = \frac{R_2}{R_1 + R_2} \quad \beta = \frac{R_2}{R_1 \cdot R_2 + R_1 \cdot R_3 + R_2 \cdot R_3}$$

$$\text{Find}(R_2, R_3) \text{ simplify} \rightarrow \begin{pmatrix} \frac{R_1 \cdot \alpha}{\alpha - 1} \\ \frac{\alpha}{\beta} - R_1 \cdot \alpha \end{pmatrix}$$

You're given full credit if you assumed  $\alpha=1$  and used one resistor in series

$$R_{th} := \frac{V_x}{\beta \cdot V_x} \rightarrow \frac{1}{\beta}$$

Full credit is also given for a dependent source and series resistor.

$$V_d := \alpha \cdot V_x \quad R_{th} := \frac{\alpha}{\beta}$$

### Problem 1 b

Let b be our reference voltage (ground). The open circuit voltage at node a is

$$V_{aoc} := V_1 \cdot \frac{R_6}{R_5 + R_6}$$

The short circuit current obtained by grounding a is

$$I_{scab} := \frac{V_1}{R_5}$$

Then

$$R_{thb} := \frac{V_{aoc}}{I_{scab}} \rightarrow \frac{R_5 \cdot R_6}{R_5 + R_6} \quad V_{thb} := V_{aoc} \rightarrow \frac{R_6 \cdot V_1}{R_5 + R_6}$$

### Problem 1 c

Ground the node in the middle. The open circuit voltage is

$$V_{aoc} := V_1 \cdot \frac{R_6}{R_5 + R_6} \quad V_{boc} := 0$$

$$V_{aboc} := V_{aoc} - V_{boc} \rightarrow \frac{R_6 \cdot V_1}{R_5 + R_6}$$

The short circuit current is

$$I_{scab} := \frac{V_{thb}}{R_{thb} + \frac{R_3 \cdot (R_1 + R_2)}{R_3 + R_1 + R_2}} \text{ simplify } \rightarrow \frac{R_6 \cdot V_1}{\left[ \frac{R_5 \cdot R_6}{R_5 + R_6} + \frac{R_3 \cdot (R_1 + R_2)}{R_1 + R_2 + R_3} \right] \cdot (R_5 + R_6)}$$

where we used the Thevenin source from part b.

$$\frac{V_{aboc}}{I_{scab}} \text{ simplify } \rightarrow \frac{R_5 \cdot R_6}{R_5 + R_6} + \frac{R_3 \cdot (R_1 + R_2)}{R_1 + R_2 + R_3}$$

The solution is more easily found by substituting the result from part b directly into the circuit of part c. Then by inspection,

$$V_{oc} := V_{thb}$$

$$R_{th} := R_{thb} + \text{par}(R_3, R_1 + R_2) \rightarrow \frac{R_5 \cdot R_6}{R_5 + R_6} + \frac{R_3 \cdot (R_1 + R_2)}{R_1 + R_2 + R_3}$$

## Problem 2

First, choose a ground. Node 4 is convenient because it gets rid of the supernode.

$$\frac{V_1 - V_x}{R_7} + \frac{V_1 - V_3}{R_4} + \frac{V_1 - V_2}{R_5} = 0$$

$$\frac{V_2 - V_1}{R_5} - \alpha \cdot V_o - I_A = 0$$

$$I_A - I_B + \frac{V_3 - V_1}{R_4} = 0$$

$V_o = V_1 - V_3$  substitute this into the second equation to get

$$\frac{V_2 - V_1}{R_5} - \alpha \cdot (V_1 - V_3) - I_A = 0$$

Rewriting the above:

$$V_1 \cdot \left( \frac{1}{R_7} + \frac{1}{R_4} + \frac{1}{R_5} \right) + V_2 \cdot \left( \frac{-1}{R_5} \right) + V_3 \cdot \left( \frac{-1}{R_4} \right) = \frac{V_x}{R_7}$$

$$V_1 \cdot \left( \frac{-1}{R_5} - \alpha \right) + V_2 \cdot \left( \frac{1}{R_5} \right) + V_3 \cdot (\alpha) = I_A$$

$$V_1 \cdot \left( \frac{-1}{R_4} \right) + V_3 \cdot \left( \frac{1}{R_4} \right) = I_B - I_A$$

### Problem 3

First, redraw the circuit so there are no overlaps.

Let  $I_{R2}$  be the current flowing through  $R_2$  consistent with the direction of  $I_y$

$$\text{KVL:} \quad I_{R2} \cdot R_2 + (I_{R2} - i_o) \cdot R_5 + \beta \cdot i_o + (I_{R2} - I_y) \cdot R_6 = 0$$

$$\text{KCL at bottom node:} \quad -I_x + i_o - I_y = 0$$

$$I_{R2}(R_2 + R_5 + R_6) + i_o(-R_5 + \beta) = I_y \cdot (R_6)$$

$$i_o = I_x + I_y$$

Note: depending on how you redrew the circuit and labeled the meshes, you may end up with several different forms of the above equations.

### Problem 4 a

Define a ground. The bottom node is convenient.

Given

$$\frac{V_x}{1\Omega} - 1A - 1A - \alpha \cdot V_R = 0$$

$$\frac{V_R}{1\Omega} + 1A = 0$$

$$V_x = 500\text{mV}$$

$$\begin{pmatrix} \alpha \\ V_x \\ V_R \end{pmatrix} := \text{Find}(\alpha, V_x, V_R) \rightarrow \begin{pmatrix} \frac{500 \cdot \text{mV} - 2 \cdot \text{A} \cdot \Omega}{\text{A} \cdot \Omega^2} \\ 500 \cdot \text{mV} \\ -\text{A} \cdot \Omega \end{pmatrix}$$

$$\alpha \rightarrow -\frac{500 \cdot \text{mV} - 2 \cdot \text{A} \cdot \Omega}{\text{A} \cdot \Omega^2} = 1.5 \cdot \text{S}$$