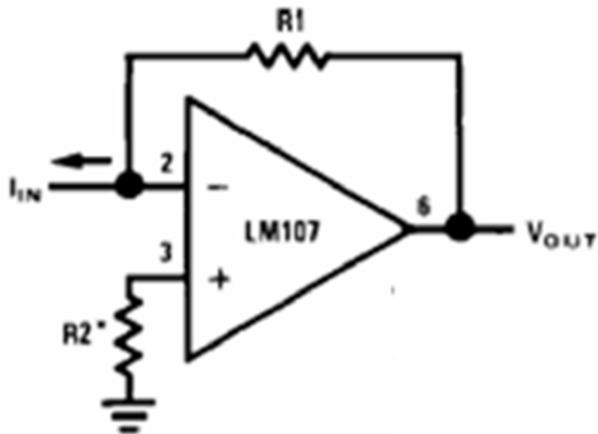


Problem 1(a)



From Golden Rule #2, no current flows into or out of terminal V_+ . Then we know the voltage drop across R_2 is zero. In turn, we know $V_+ = 0$.

From Golden Rule #1, $V_- = V_+ = 0$.

$$\text{KCL @ } V_-: -I_{IN} + 0 + \frac{(V_{out}-0)}{R_1} = 0$$

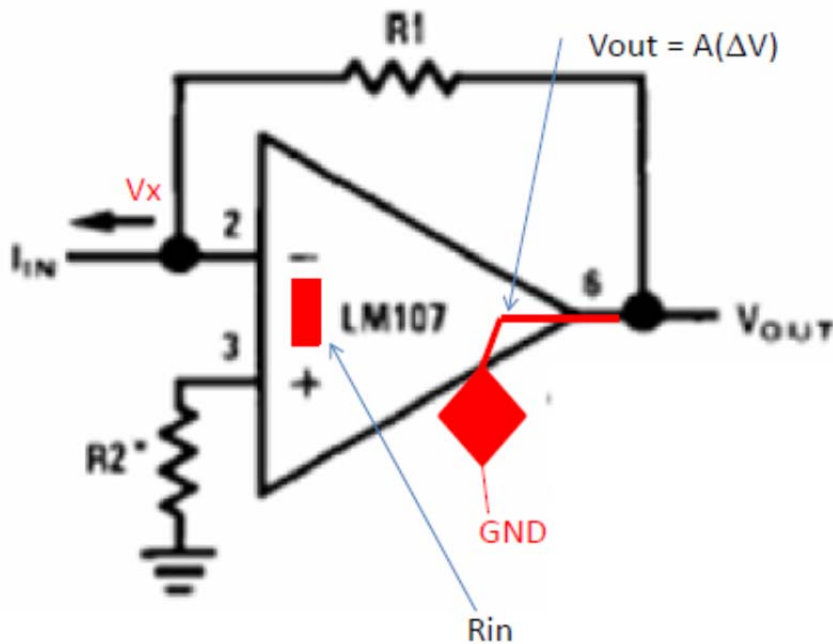
Then, we know: $V_{out} = R_1 * I_{IN}$

Problem 1(b):

No. When R_2 is a short circuit, $V_+ = 0$. By Golden Rule #1, $V_- = V_+ = 0$. The KCL equation at V_- is still the same. Therefore, the answer does not change.

Problem 1(c):

Considering the non-ideality, we can redraw the circuit:



1. From the amplifier: $V_{out} = A(\Delta V) = A (V_p - V_n)$.

2. KCL @ V_x : $-I_{IN} + \frac{(0 - V_n)}{R_2 + R_{in}} + \frac{V_{out} - V_n}{R_1} = 0$

3. From voltage division, $\Delta V = V_p - V_n = V_n \cdot \frac{-R_{in}}{R_{in} + R_2}$

Solving the above system:

$$V_x = \frac{-I_{IN}}{\frac{A R_{in}}{R_1(R_2 + R_{in})} + \frac{1}{R_1} + \frac{1}{R_2 + R_{in}}} = \frac{-I_{IN}(R_2 + R_{in})R_1}{A R_{in} + R_1 + (R_2 + R_{in})}$$

$$V_{out} = A \cdot \Delta V = A \cdot -V_n \frac{R_{in}}{R_{in} + R_2} = \frac{A I_{IN} R_1 R_{in}}{A R_{in} + R_1 + (R_2 + R_{in})}$$

Problem 1(d)

Conceptually, to reduce the effect of R_{in} on V_{out} , we should try to reduce the current flowing through R_{in} . Therefore, we should have as **LARGE** R_2 as possible.

Mathematically:

From 1(c):

$$V_{out} = \frac{A I_{IN} R_1 R_{in}}{A R_{in} - R_1 - (R_2 + R_{in})}$$

The effect of R_{in} on V_{out} =

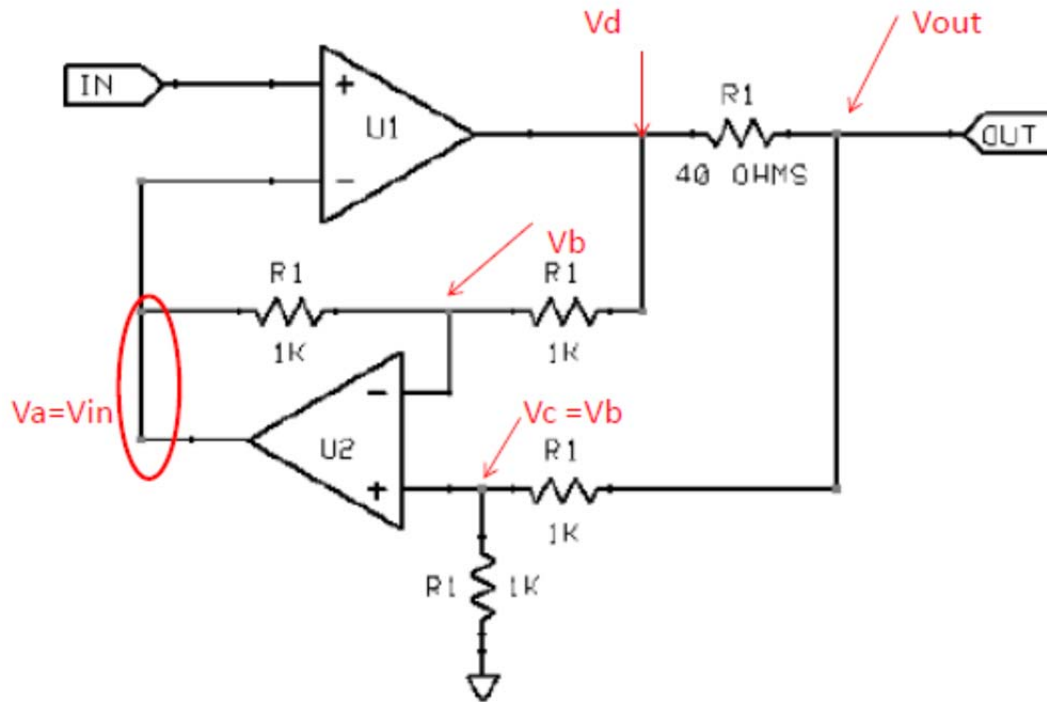
$$\begin{aligned} \frac{\partial V_{out}}{\partial R_{in}} &= \frac{-A I_{IN} R_1 \{R_1 + R_2\}}{[A R_{in} - R_1 - (R_2 + R_{in})]^2} \\ &= \frac{O(R_2)}{O(R_2^2)} \end{aligned}$$

(Big O notation, the numerator has R_2 in the first order, and the denominator has R_2 in the second order.

Therefore, when $R_2 \rightarrow \infty$, $\frac{\partial V_{out}}{\partial R_{in}} = 0$ (changes of R_{in} has no effect on V_{out})

Therefore, we should have as **LARGE** R_2 as possible.

Problem 2(a)



$$1) \text{ KCL @ } V_b: \frac{V_{in}-V_b}{1k} + \frac{V_d-V_b}{1k} + 0 = 0$$

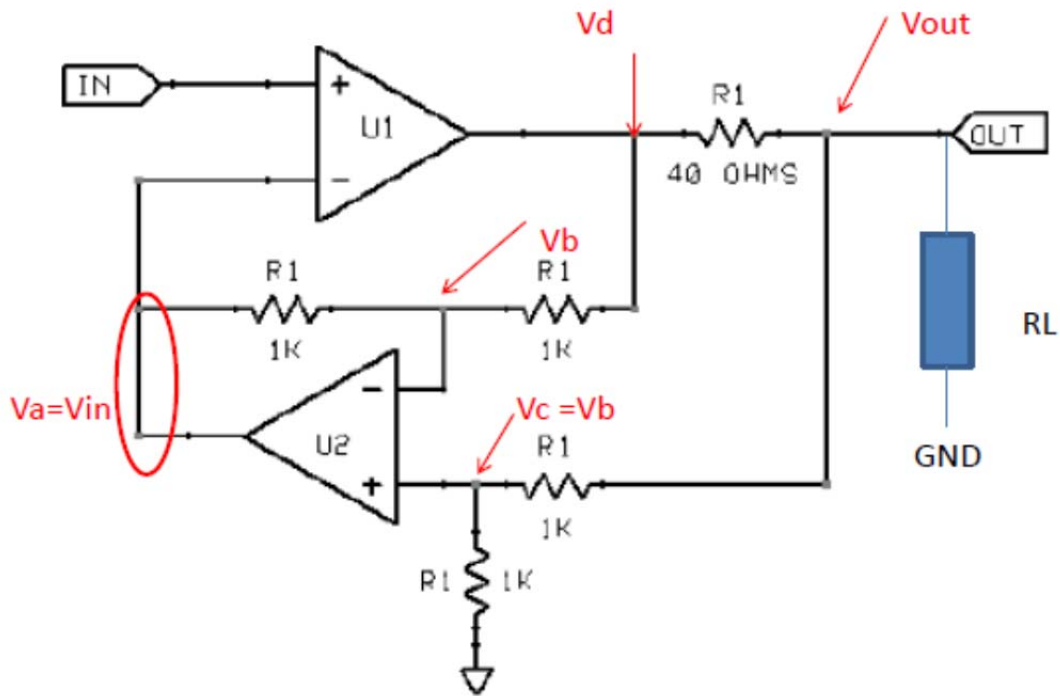
$$2) \text{ KCL @ } V_c: 0 + \frac{0-V_b}{1k} + \frac{V_{out}-V_b}{1k} = 0$$

$$3) \text{ KCL @ } V_{out}: \frac{V_d-V_{out}}{40} + \frac{V_b-V_{out}}{1k} = 0$$

Three unknowns: V_b , V_d , and V_{out} ; three equations.

After solving for the system, we get $V_{out} = -50V_{in}$

Problem 2(b)



$$1) \text{ KCL @ } V_b: \frac{V_{in} - V_b}{1k} + \frac{V_d - V_b}{1k} + 0 = 0$$

$$2) \text{ KCL @ } V_c: 0 + \frac{0 - V_b}{1k} + \frac{V_{out} - V_b}{1k} = 0$$

$$3) \text{ KCL @ } V_{out}: \frac{V_d - V_{out}}{40} + \frac{V_b - V_{out}}{1k} + \frac{0 - V_{out}}{R_L} = 0$$

Three unknowns: V_b , V_d , and V_{out} ; three equations.

After solving for the system, we get $V_{out} = V_{in} \cdot \frac{-50}{1 + \frac{2000}{R_L}}$

$$\text{Then, } I_L = V_{out} / R_L = V_{in} \cdot \frac{-50}{R_L + 2000}$$

Problem 3(a)

We want to be able to change the gain without changing the internal design of the op-amp. In addition, negative feedback provides stable gain as well.

Problem 3(b)

$\text{Amp} * 100 = 15\text{V}$. $\text{Amp, max} = 0.15\text{V}$. The maximum amplitude without clipping is 0.15V .

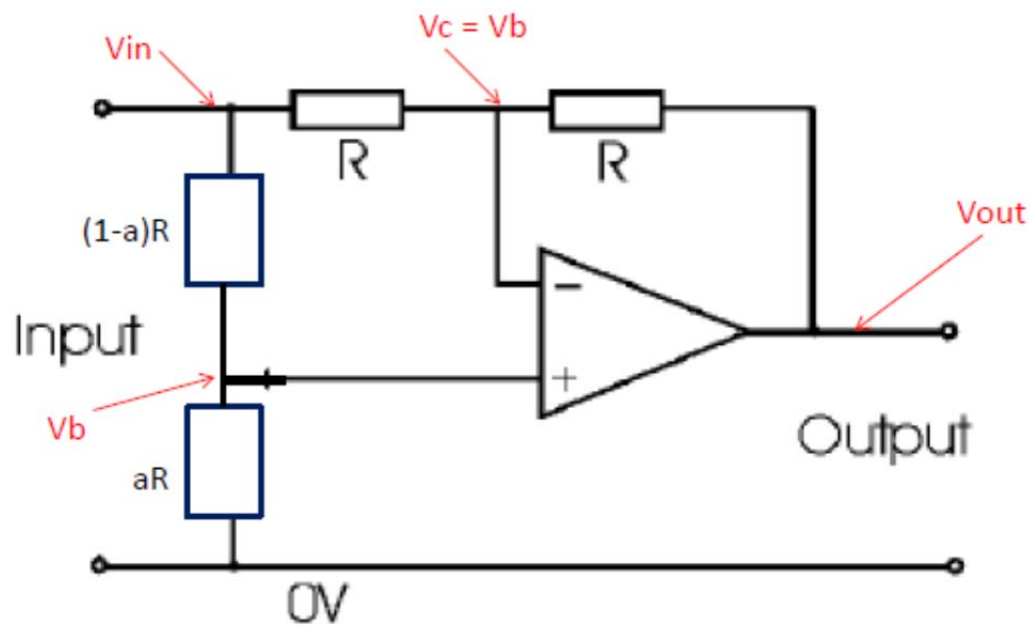
Problem 3(c)

When clipping occurs, a pure sinusoidal wave is no longer sinusoidal. The distorted wave now contains waves of some high frequencies (since such clipped waves can be modeled as a combination of numerous sinusoids of multiple frequencies). Because of these additional frequencies, the sound is no longer the same.

Problem 3(d)

$\text{dB} = 20 \log (\text{Gain}) = 20 \log (100) = 40 \text{ dB}$.

Problem 4



0) Label all the nodes as shown above.

1) From Golden Rule #1, $V_c = V_b$.

2) KCL @ V_b : $\frac{V_{in}-V_b}{(1-a)R} + 0 + \frac{0-V_b}{aR} = 0$

3) KCL @ V_c : $\frac{V_{in}-V_b}{R} + 0 + \frac{V_{out}-V_b}{R} = 0$

From the two KCL equations, We can solve for V_{out} .

$$V_b = aV_{in}$$

$$V_{out} = (2a - 1)V_{in}$$

$$\frac{V_{out}}{V_{in}} = 2a - 1$$
