

**EECS 40, Fall 2006**  
**Prof. Chang-Hasnain**  
**Midterm #2**

October 25, 2006

Total Time Allotted: 50 minutes

Total Points: 100 / Bonus: 10 pts

1. This is a closed book exam. However, you are allowed to bring one page (8.5" x 11"), single-sided notes PLUS your 1-page notes from midterm 1.
2. No electronic devices, i.e. calculators, cell phones, computers, etc.
3. Slide rules are allowed.
4. SHOW all the steps on the exam. **Answers without steps will be given only a small percentage of credits.** Partial credits will be given if you have proper steps but no final answers.
5. **Remember to put down units.** Points will be taken off for answers without units.

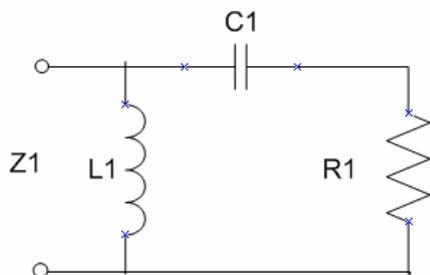
Last (Family) Name: Perfect

First Name: Peter

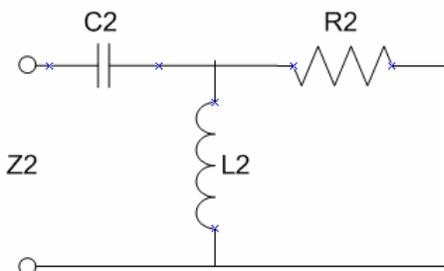
Student ID: 314159265 Discussion Session: 2718

Signature: PP

<b>Score:</b>	<b>110</b>
Problem 1 (16 pts) Complex Impedances	16
Problem 2 (54 pts): Bode Plots	54
Bonus (10 pts):	10
Problem 3 (30 pts): Second-order Circuits	30
Total	110

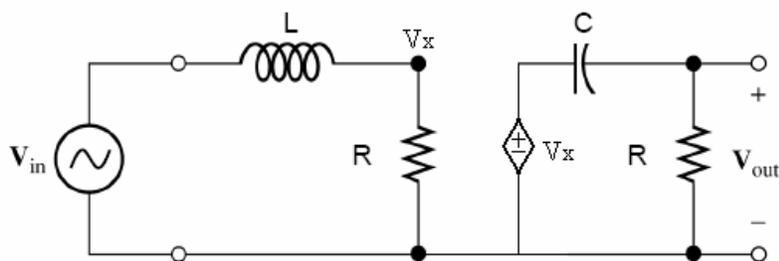
**1. [16 points] Parallel and Series Complex Impedance**a) [8 pts] What is the complex impedance  $Z_1$ ?

$$\begin{aligned}
 Z_1 &= Z_L \parallel (Z_C + R) \\
 &= j\omega L \parallel \left( \frac{1}{j\omega C} + R \right) = j\omega L \parallel \left( \frac{1 + j\omega RC}{j\omega C} \right) \\
 &= \left( \frac{1}{j\omega L} + \frac{j\omega C}{1 + j\omega RC} \right)^{-1} = \left( \frac{1 + j\omega RC + (j\omega L)(j\omega C)}{j\omega L(1 + j\omega RC)} \right)^{-1} = \left( \frac{j\omega L(1 + j\omega RC)}{1 + j\omega RC + j^2 \omega^2 LC} \right) \\
 &= \frac{-\omega^2 RLC + j\omega L}{1 - \omega^2 LC + j\omega RC}
 \end{aligned}$$

*(subscripts omitted for clarity)*b) [8 pts] What is the complex impedance  $Z_2$ ?

$$\begin{aligned}
 Z_2 &= Z_C + (Z_L \parallel R) \\
 &= \frac{1}{j\omega C} + (j\omega L \parallel R) = \frac{1}{j\omega C} + \left( \frac{1}{j\omega L} + \frac{1}{R} \right)^{-1} \\
 &= \frac{1}{j\omega C} + \left( \frac{j\omega RL}{R + j\omega L} \right) = \left( \frac{R + j\omega L + (j\omega C)(j\omega RL)}{j\omega C(R + j\omega L)} \right) = \left( \frac{R + j\omega L + j^2 \omega^2 RLC}{j\omega RC + j^2 \omega^2 LC} \right) \\
 &= \frac{R - \omega^2 RLC + j\omega L}{- \omega^2 LC + j\omega RC}
 \end{aligned}$$

## 2. [54 points] Bode Plots:



(a) [10 points] For the above circuit, show 
$$H(f) = \frac{1}{1 + j\frac{f}{f_2}} \times \frac{1}{1 - j\frac{f_1}{f}}$$

Express  $f_1$  and  $f_2$  in terms of  $R$ ,  $L$ ,  $C$ . (Hint: Remember  $\omega = 2\pi f$ )

Voltage divider on left:

$$V_x = \frac{R}{R + j\omega L} V_{in} = \frac{1}{1 + j\omega \frac{L}{R}} V_{in}$$

Voltage divider on right:

$$V_{out} = \frac{R}{R + \frac{1}{j\omega C}} V_x = \frac{1}{1 - j\frac{1}{\omega RC}} V_x$$

Combining, 
$$V_{out} = \frac{1}{1 - j\frac{1}{\omega RC}} \times \frac{1}{1 + j\omega \frac{L}{R}} V_{in}$$

Swapping the terms, 
$$H(f) = \frac{V_{out}}{V_{in}} = \frac{1}{1 + j\omega \frac{L}{R}} \times \frac{1}{1 - j\frac{1}{\omega RC}}$$

$$= \frac{1}{1 + j2\pi f \frac{L}{R}} \times \frac{1}{1 - j\frac{1}{2\pi f RC}}$$

$f_1$  occurs with  $f$  in the denominator, and  $f_2$  with  $f$  in the numerator:

$$f_1 = \frac{1}{2\pi RC} \qquad f_2 = \frac{R}{2\pi L}$$

(b) [6 points] Now Let  $R = 1\text{ k}\Omega$ ,  $L = 0.16\text{ mH}$ ,  $C = 0.16\text{ }\mu\text{F}$ , what are  $f_1$  and  $f_2$ ? Remember to put down units.

$$f_1 = \frac{1}{2\pi(1\text{ k}\Omega)(0.16\text{ }\mu\text{F})} = \frac{6.25}{2\pi \times 10^{-3}\text{ s}} \approx 1\text{ kHz}$$

$$f_2 = \frac{1\text{ k}\Omega}{2\pi(0.16\text{ mH})} = \frac{6.25 \times 10^6}{2\pi\text{ s}} \approx 1\text{ MHz}$$

(c) [22 pt] Bode Magnitude Plot. **You must put down all the steps leading to your results.**

**Hint:** You may consider  $f_1 \ll f_2$

[4 points] Write down the expression for  $y = 10\log|H(f)|^2$

$$\begin{aligned} y &= 10\log\left|\frac{1}{1+j\frac{f}{f_2}}\right|^2 + 10\log\left|\frac{1}{1-j\frac{f_1}{f}}\right|^2 \\ &= -10\log\left(1+\left(\frac{f}{f_2}\right)^2\right) - 10\log\left(1+\left(\frac{f_1}{f}\right)^2\right) \end{aligned}$$

Units are: dB

Note: The other acceptable expression can be found by multiplying the terms in  $H(f)$ , and finding the magnitude of the resulting product.

[4 points] As frequency goes to a very small value, what is the slope of  $y$  as a function of  $\log f$ ?

Constant 1 dominates in left term,  $f^{-1}$  dominates in the right term:

$$\begin{aligned} y &= -10\log(1) - 10\log\left(\frac{f_1}{f}\right)^2 = 0 + 20\log\left(\frac{f}{f_1}\right) \\ &= 20\log f - 20\log f_1 \end{aligned}$$

Slope: 20 dB / decade

[4 points] As frequency goes to a very large value, what is the slope of  $y$  as a function of  $\log f$ ?

$f$  dominates in left term, constant 1 dominates in the right term:

$$\begin{aligned} y &= -10\log\left(\frac{f}{f_2}\right)^2 - 10\log(1) = -20\log\left(\frac{f}{f_2}\right) - 0 \\ &= -20\log f - (-20\log f_2) \end{aligned}$$

Slope: -20 dB / decade

[4 points] What is  $y$ ,  $f_1 \ll f \ll f_2$ ?

In both terms, constant 1 dominates:

$$\begin{aligned} y &= -10\log(1) - 10\log(1) = -0 - 0 \\ &= 0 \text{ dB} \end{aligned}$$

[2 points] What is  $y$  at  $f_1$ ?

$$y = -10\log\left(1+\left(\frac{f_1}{f_2}\right)^2\right) - 10\log\left(1+\left(\frac{f_1}{f_1}\right)^2\right)$$

$$\begin{aligned} \text{Since } f_1 \ll f_2, \quad y &= -10\log(1) - 10\log(1+1) \\ &= -10\log 2 = -3 \text{ dB} \end{aligned}$$

[2 points] What is  $y$  at  $f_2$  ?

$$y = -10 \log \left( 1 + \left( \frac{f_2}{f_2} \right)^2 \right) - 10 \log \left( 1 + \left( \frac{f_1}{f_2} \right)^2 \right)$$

$$\begin{aligned} \text{Again, } f_1 \ll f_2, y &= -10 \log(1+1) - 10 \log(1) \\ &= -10 \log 2 = -3 \text{ dB} \end{aligned}$$

[2 points] What filter is this?

*Bandpass filter*

**Bonus [5 points]** If the input  $|V_{in}| = 1 \text{ V}$  and the frequency is 1 MHz, what is the output  $|V_{out}|$  ?

$$f_2 = 1 \text{ MHz, so } y = -3 \text{ dB} = 10 \log\left(\frac{1}{2}\right)$$

$$\frac{1}{2} = |H(f)|^2 = \left| \frac{V_{out}}{V_{in}} \right|^2 \qquad |V_{out}| = \frac{1}{\sqrt{2}} |V_{in}| \approx 0.707 \text{ V}$$

**Bonus [5 points]** If the input  $|V_{in}| = 1 \text{ V}$  and the frequency is 10 MHz, what is the output  $|V_{out}|$  ?

$$10 \text{ MHz} = 10 f_2 : \text{one decade past break frequency}$$

*At large  $f$ , slope is -20 dB/decade. Thus,  $y = -20 \text{ dB}$ , since  $f$  is 1 decade higher than the break frequency, where the Bode approximation is 0 dB.*

$$y = -20 \text{ dB} = 10 \log\left(\frac{1}{100}\right)$$

$$\frac{1}{100} = |H(f)|^2 = \left| \frac{V_{out}}{V_{in}} \right|^2 \qquad |V_{out}| = \frac{1}{\sqrt{100}} |V_{in}| = 0.1 \text{ V}$$

**(d) [16 pt total] Bode Phase Plot. You must put down all the steps leading to your results. Hint: You may consider  $f_1 \ll f_2$**

[4 points] Write down the expression for  $\angle H(f)$

$$\begin{aligned} \angle H(f) &= \tan^{-1} 0 - \tan^{-1} \left( \frac{f}{f_2} \right) + \tan^{-1} 0 - \tan^{-1} \left( -\frac{f_1}{f} \right) \\ &= \tan^{-1} \left( \frac{f_1}{f} \right) - \tan^{-1} \left( \frac{f}{f_2} \right) \end{aligned}$$

[4 points] What does the value of  $\angle H(f)$  approaches to as  $f \rightarrow 0$  ?

$$\angle H(f) = \tan^{-1}(\infty) - \tan^{-1}(0) = \frac{\pi}{2} \text{ radians}$$

[4 points] What does the value of  $\angle H(f)$  approaches to as  $f \rightarrow \infty$ ?

$$\angle H(f) = \tan^{-1}(0) - \tan^{-1}(\infty) = -\frac{\pi}{2} \text{ radians}$$

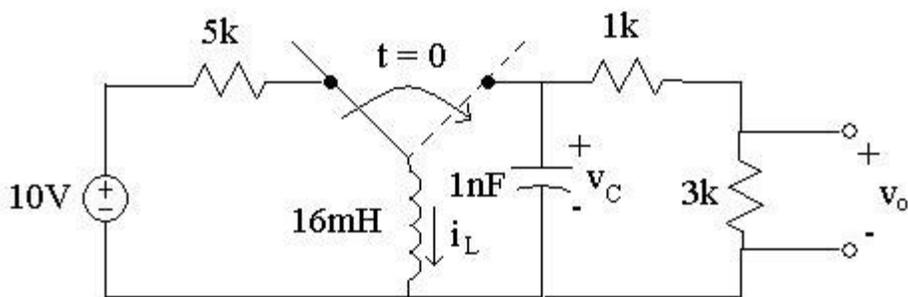
[2 points] What is  $\angle H(f)$  at  $f = f_1$ ?

$$\begin{aligned} \angle H(f) &= \tan^{-1}\left(\frac{f_1}{f_1}\right) - \tan^{-1}\left(\frac{f_1}{f_2}\right) = \tan^{-1}(1) - \tan^{-1}(0) \\ &= \frac{\pi}{4} \text{ radians} \end{aligned}$$

[2 points] What is  $\angle H(f)$  at  $f = f_2$ ?

$$\begin{aligned} \angle H(f) &= \tan^{-1}\left(\frac{f_1}{f_2}\right) - \tan^{-1}\left(\frac{f_1}{f_2}\right) = \tan^{-1}(0) - \tan^{-1}(1) \\ &= -\frac{\pi}{4} \text{ radians} \end{aligned}$$

## 3. [30 points] Second-order Circuits:



Assume the switch has been to the left for a long time before switching to the right at  $t = 0$ .

(a) Find the following values: [18 points] (Hint: What is  $v_o(t)$  in terms of  $v_c(t)$ ?)

$i_L(0+) = \frac{10\text{ V}}{5\text{ k}\Omega} = 2\text{ mA}$ <p><i>Inductor is short circuit right before switch is thrown.</i></p>	$i_L(\infty) = 0\text{ A}$ <p><i>All power is dissipated through resistors.</i></p>
$v_c(0+) = 0\text{ V}$ <p><i>No stored charge.</i></p>	$v_c(\infty) = 0\text{ V}$ <p><i>Discharges through resistors.</i></p>
$v_o(0+) = 0\text{ V}$ <p><i>All current initially flows through capacitor (short).</i></p>	$v_o(\infty) = 0\text{ V}$ <p><i>Power dissipated, no current.</i></p>
$\frac{d}{dt} i_L(0+) = 0\text{ A/s}$ $v_L(0+) = L \frac{d}{dt} i_L(0+) = V_c(0+) = 0\text{ V}$	
$\frac{d}{dt} v_c(0+) = -2\text{ MV/s}$ $i_c(0+) = C \frac{d}{dt} v_c(0+) \Rightarrow \frac{d}{dt} v_c(0+) = \frac{-i_L(0+)}{C} = \frac{-2\text{ mA}}{1\text{ nF}} = -2 \times 10^6\text{ V/s}$	
$\frac{d}{dt} v_o(0+) = -1.5\text{ MV/s}$ <p><i>Voltage divider: <math>v_o = \frac{3\text{ k}\Omega}{1\text{ k}\Omega + 4\text{ k}\Omega} v_c \Rightarrow \frac{d}{dt} v_o(t) = \frac{3}{4} \frac{d}{dt} v_c(t)</math></i></p>	

(b) [6 points] Write the differential equation in terms of  $v_c$ .

*KCL @ node above capacitor, (the 2 resistors are combined into R):*

$$\begin{aligned} 0 &= i_L(t) + C \frac{d}{dt} v_c(t) + \frac{1}{R} v_c(t) \\ &= \frac{1}{L} \int v_c(t) dt + C \frac{d}{dt} v_c(t) + \frac{1}{R} v_c(t) \end{aligned}$$

*Differentiating,*

$$\begin{aligned} 0 &= \frac{1}{L} v_c(t) + C \frac{d^2}{dt^2} v_c(t) + \frac{1}{R} \frac{d}{dt} v_c(t) \\ 0 &= \frac{d^2}{dt^2} v_c(t) + \frac{1}{RC} \frac{d}{dt} v_c(t) + \frac{1}{LC} v_c(t) \end{aligned}$$

(c) [6 points] What are the values of the natural frequency ( $\omega_0$ ) and the damping ratio ( $\zeta$ )?

*Damped harmonic oscillation:*

$$0 = \frac{d^2}{dt^2} v_c(t) + 2\alpha \frac{d}{dt} v_c(t) + \omega_0^2 v_c(t)$$

$$\begin{aligned} \omega_0 &= \sqrt{\frac{1}{LC}} = \sqrt{\frac{1}{(16 \text{ mH})(1 \text{ nF})}} = \frac{1}{\sqrt{16 \times 10^{-12} \text{ s}^2}} \\ &= 250 \text{ krad / s} \end{aligned}$$

$$\begin{aligned} \alpha &= \frac{1}{2RC} = \frac{1}{2(4 \text{ k}\Omega)(1 \text{ nF})} = \frac{1}{8 \times 10^{-6} \text{ s}} \\ &= 125 \text{ krad / s} \end{aligned}$$

$$\zeta = \frac{\alpha}{\omega_0} = \frac{125 \text{ krad / s}}{250 \text{ krad / s}} = 0.5 \quad (\text{underdamped behavior})$$